

Impulse Response Measurements in the Presence of Clock Drift

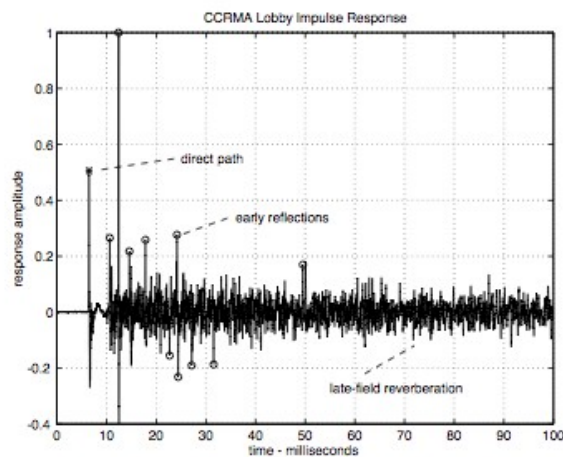
Nicholas J. Bryan, Miriam A. Kolar, and Jonathan S. Abel

Stanford University | CCRMA



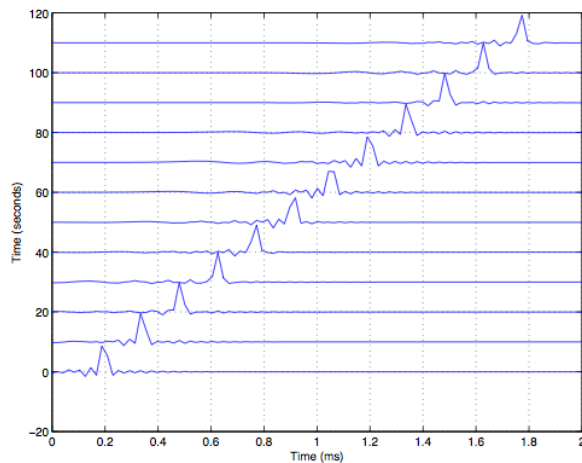
Introduction + Motivation

- Characterize the acoustic properties of a room in the form of an impulse response (IR)
- Occasionally difficult to record and playback on a single device as found in archeological acoustics (Chavín de Huántar—Miriam Kolar, et al.)

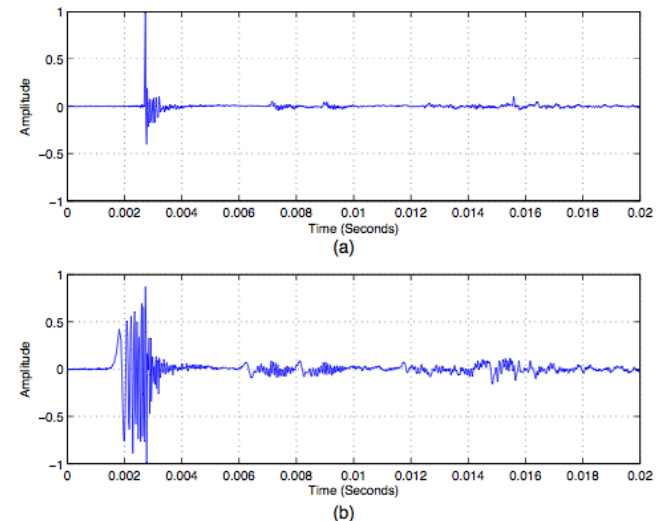


IR Measurements w/Clock Drift

- Different devices results in different digital clocks causing small timing differences in playback and record signals
- The misalignment accumulates over time for robust impulse response measurement techniques using convolution/correlation



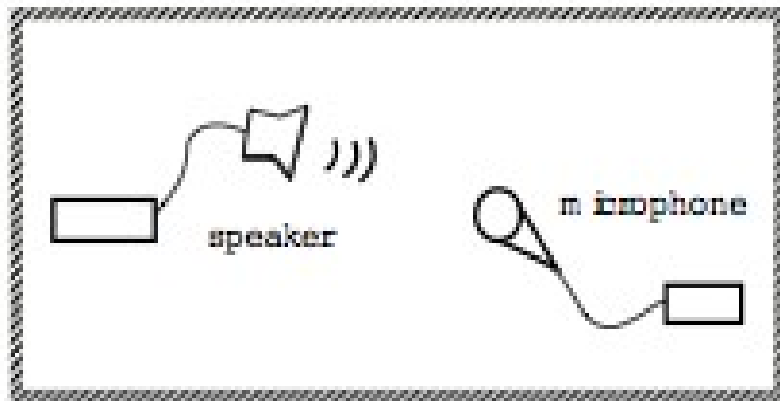
Recorded impulse train with clock drift
over two minutes, drift 1-2 ms / minute



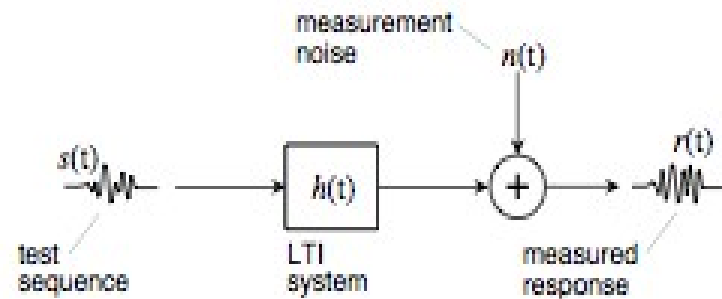
Measurement with and without clock drift

Impulse Response Measurements

- Measurement model $r(t) = s(t) * h(t) + n(t)$



Measurement system setup



Measurement model

Convolution IR Measurement Methods

Cyclic convolution

- Sinusoidal sweeps
 - linear, exponential, etc.
- Maximum length sequences
- Allpass chirps

Acyclic convolution

- **Sinusoidal sweeps**
 - linear, exponential, etc.
- Golay codes
 - binary, ternary, etc.

Sinusoidal Sweeps

- Increasingly popular, straightforward implementation
- Robust measurements against weak non-linearities
- Offer thorough theoretical analysis and alternative methods in the presence of clock drift

Acyclic IR Measurements

- Given the measurement model $r(t) = s(t) * h(t) + n(t)$ we assume, there exist an inverse filter $f(t)$ such that $s(t) * f(t) \approx \delta(t)$
- The from the measurement model we can recover the impulse response via

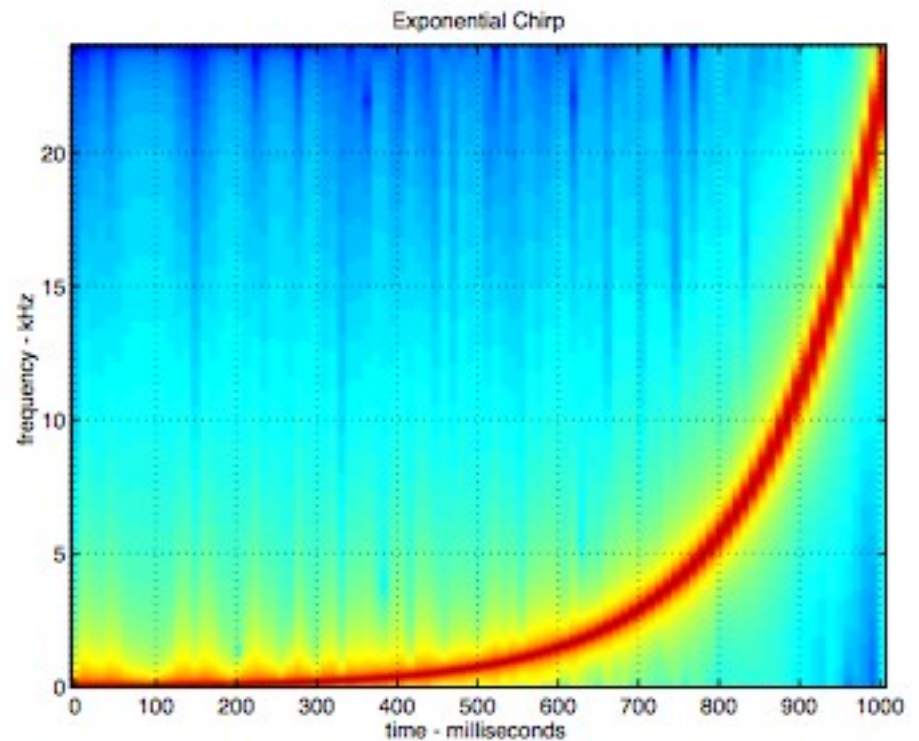
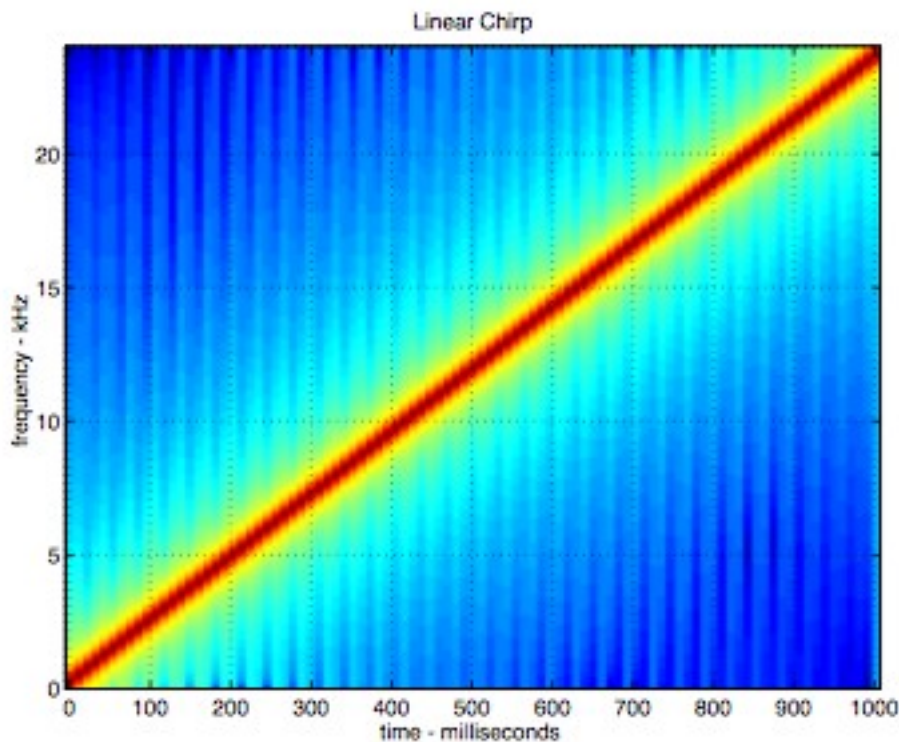
$$\begin{aligned} h(t) &= r(t) * f(t) \\ &= h(t) * s(t) * f(t) + n(t) * f(t) \\ &\approx h(t) \end{aligned}$$

Sinusoidal Sweeps

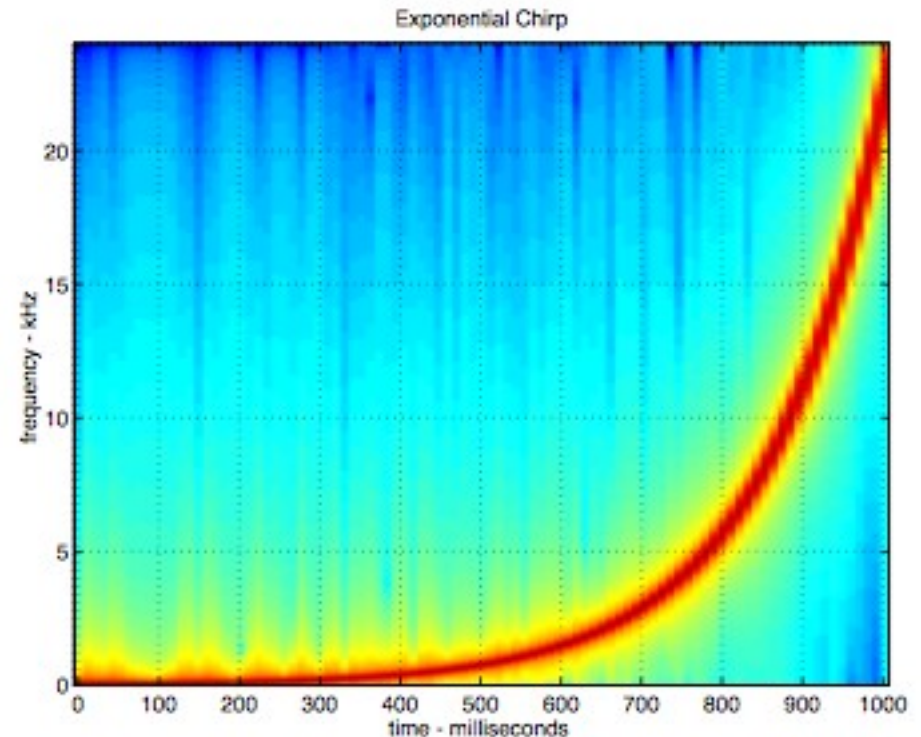
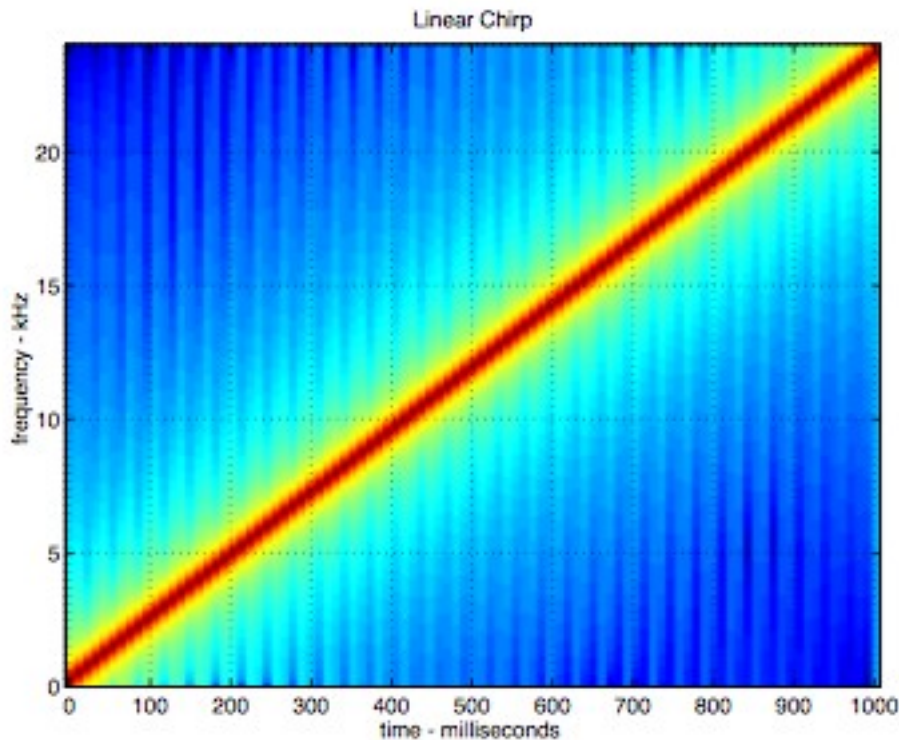
$$s(t) = \sin \phi(t) \quad \phi(t) = \int_0^t \omega(\nu) d\nu$$

$$\omega_{lin}(t) = \left(\frac{\omega_1 - \omega_0}{\tau} \right) t + \omega_0$$

$$\omega_{exp}(t) = \omega_0 \exp \left\{ \frac{t}{\tau} \ln \omega_1 / \omega_0 \right\}$$



Sine Sweep Group Delay



- For smooth phase, the group delay is the time delay of the amplitude envelope of a sinusoid
- Functional inverse of frequency trajectory

Sinusoidal Sweep Transforms

- The frequency response $S(\omega)$ can be formulated via a magnitude and phase decomposition $|S(\omega)|e^{j\phi(\omega)}$.

$$|S(\omega)| \approx 1 / \sqrt{\frac{1}{2} \left| \frac{d\gamma(\omega)}{d\omega} \right|} \quad \phi_s(\omega) = - \int_0^\omega \gamma(\nu) d\nu$$

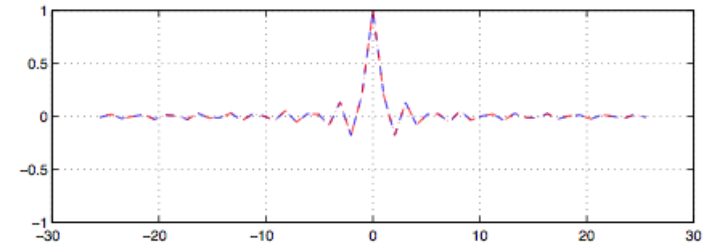
where the group delay $\gamma(\omega)$ is the functional inverse of $\omega(t)$ (Abel 2004)

Sinusoidal Sweep Inverse Filter

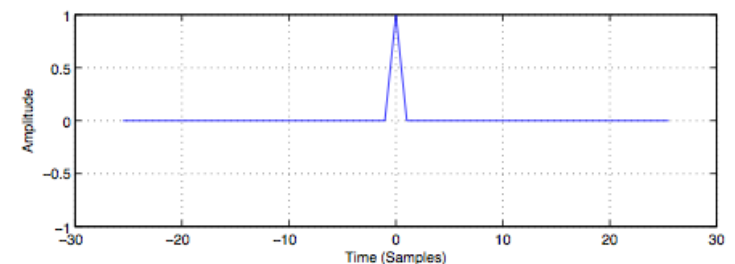
- $f(t)$ can be constructed using various methods
 - Time-reversal plus whitening filter
 - Numerical inversion via $\text{IFFT}(1/\text{FFT}(s(t)))$
 - **Closed form approximation**

$$f(t) \approx \left| \frac{d\omega(-t)}{dt} \right| s(-t)$$

$$|F(\omega)| \approx \sqrt{\frac{1}{2} \left| \frac{d\gamma(\omega)}{d\omega} \right|} \quad \phi_f(\omega) = \int_0^\omega \gamma(\nu) d\nu$$



(a)



(b)

Clock Drift Analysis

- For convenience, a single clock can be chosen as reference with the other as drifting resulting in two scenarios
 - drifting playback clock
 - drifting record clock

$$\begin{aligned} \cancel{h(t)} &= \cancel{r(t)} * f(t) \\ &= \cancel{h(t)} * \cancel{s(t)} * f(t) \end{aligned}$$

IR deconvolution process is corrupted

Drifting Playback Clock

- Recorded signal $\tilde{r}(t) = h(t) * s(\alpha_p t)$
- Propagates through to the impulse response convolution process via

$$\begin{aligned}\tilde{h}(t) &= \tilde{r}(t) * f(t) \\ &= h(t) * (s(\alpha_p t) * f(t)) \\ &= h(t) * d_p(t)\end{aligned}$$

- IR is filtered by a drift filter

$$d_p(t) = s(\alpha_p t) * f(t)$$

Drifting Record Clock

- Recorded signal $\tilde{r}(t) = h(\alpha_r t) * s(\alpha_r t)$
- Propagates through to the impulse response convolution process via

$$\begin{aligned}\tilde{h}(t) &= \tilde{r}(t) * f(t) \\ &= h(\alpha_r t) * (s(\alpha_r t) * f(t))\end{aligned}$$

- IR is filtered by a drift filter $d_r(t) = s(\alpha_r t) * f(t)$ and resampled
- Resampling is typically negligible

Sine Sweep Clock Drift

- A stretch α in the time scale is equivalent to a stretch in the frequency trajectory and hence the group delay

$$\begin{aligned}\tilde{\gamma}(\omega) &\approx \alpha \gamma(\omega) \\ &\approx (1 + \epsilon) \gamma(\omega)\end{aligned}$$

- As before, the magnitude and phase of a sweep can be solely expressed as a function of the group delay

Sine Sweep Transform w/Drift

- Modify the magnitude and phase of a sine sweep with drifting clocks resulting

$$|\tilde{S}(\omega)| \approx \sqrt{1 + \epsilon} |S(\omega)| \quad \tilde{\phi}_s(\omega) = -(1 + \epsilon) \phi(\omega)$$

- The drift filter is then

$$\begin{aligned} D(\omega) &\approx \sqrt{1 + \epsilon} \exp\{-j \epsilon \phi_s(\omega)\} \\ &\approx \exp\{-j \epsilon \phi_s(\omega)\} \end{aligned}$$

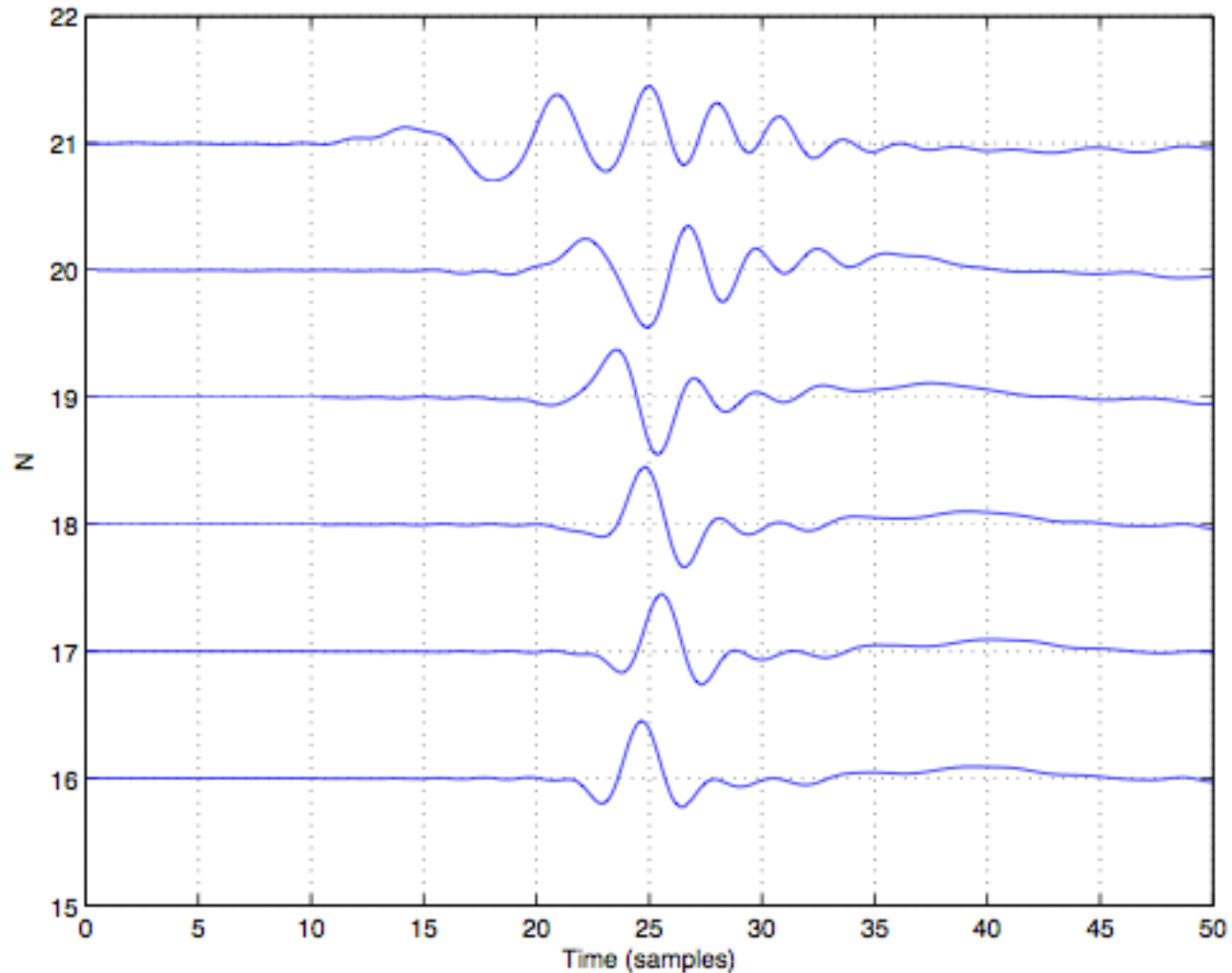
Sine Sweep Transform w/Drift

- The drift filter is an allpass filter for sine sweeps
 - Frequency trajectory of the same type as input sweep
 - Function of drift rate and sine sweep length

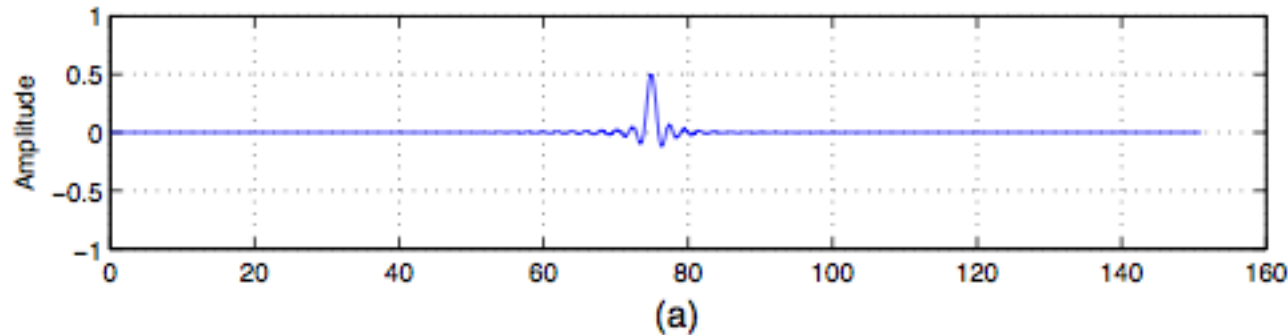
$$\begin{aligned} D(\omega) &\approx \sqrt{1 + \epsilon} \exp\{-j \epsilon \phi_s(\omega)\} \\ &\approx \exp\{-j \epsilon \phi_s(\omega)\} \end{aligned}$$

Dependent on Sine Sweep Length

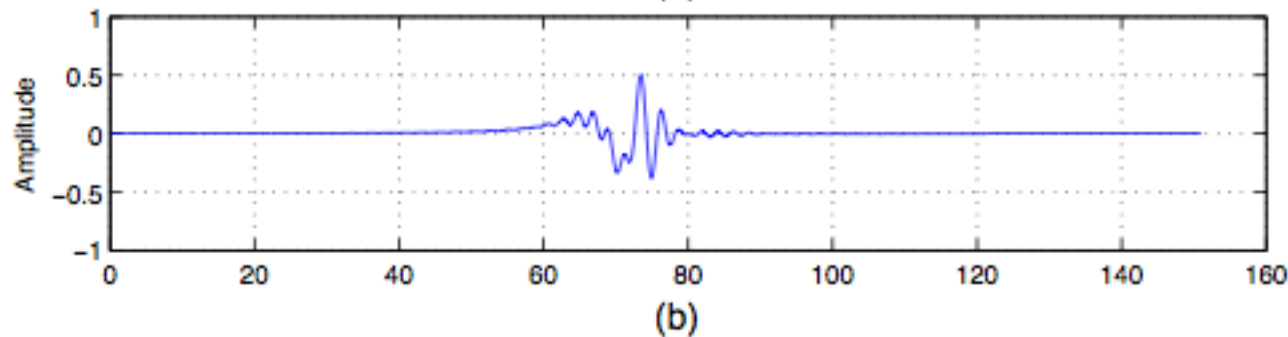
Sweep length
in samples = 2^N



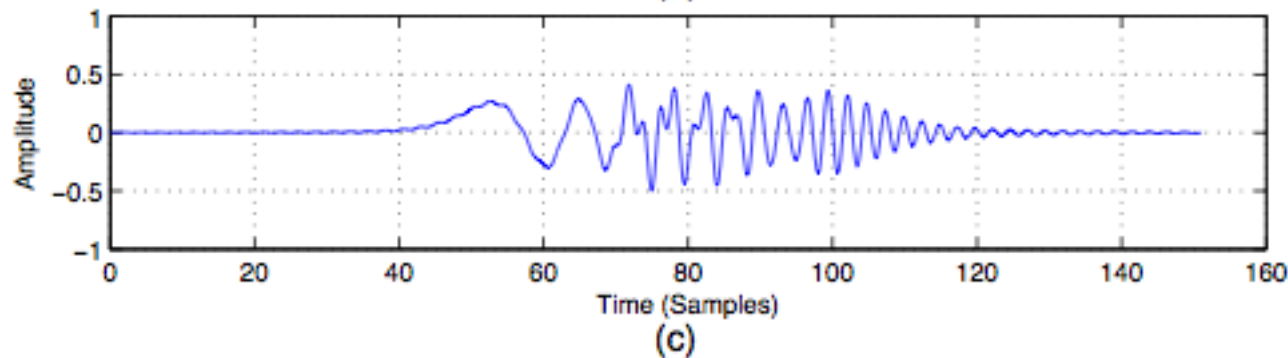
Dependent on Frequency Trajectory



No drift



Exponential trajectory
w/drift



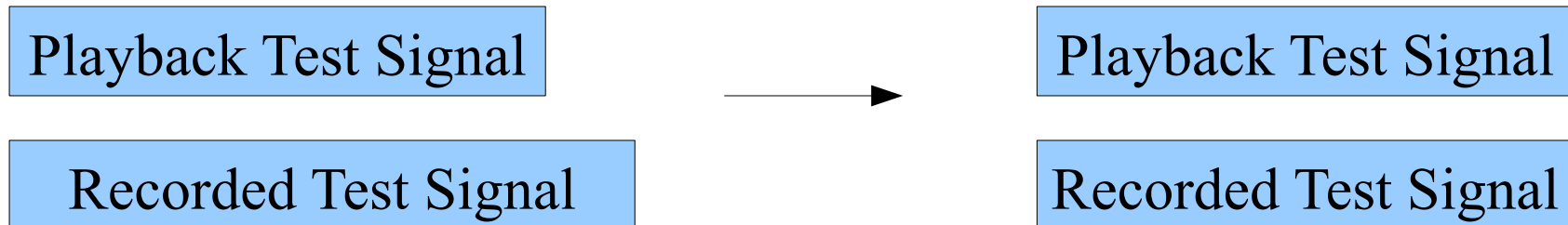
Linear trajectory
w/drift

Clock Drift Compensation

- Desirable to remove drift effects via post-processing
 - Resampling
 - Compensation Filtering

Resampling Compensation

- Resample the recorded test signal or inverse filter prior to convolution
- Applicable to all convolution-based IR methods



Compensation Filtering


- Apply a compensation filter after convolution
- Applicable to sinusoidal sweeps
- Applied drift filter is allpass, so the compensation filter is the time-flip



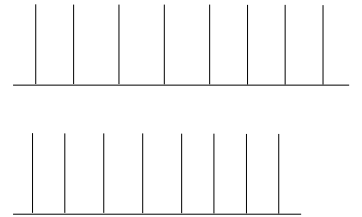
Clock Drift Estimation

- Both compensation methods need an estimate of the clock drift
- Explicit Estimation
 - Direct electronic loopback recording of an periodic impulse train, noting the recorded time indices
- Implicit Estimation
 - Direct electronic loopback of sine sweep or inverse

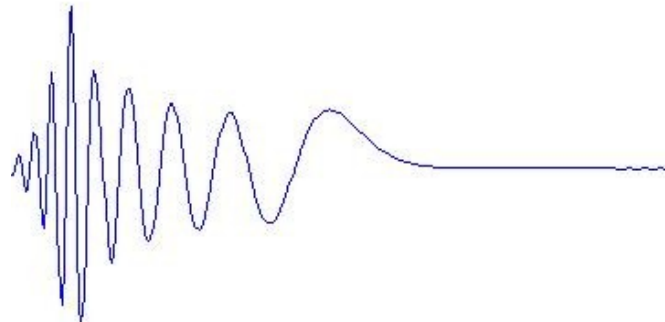
Explicit Estimation

- Record high frequency impulse train 
- Note time indices of peaks and compute least-squares estimate of the drift rate α via

$$\min_{\alpha} \| \mathbf{t} - \mathbf{t}_{meas}\alpha \|$$

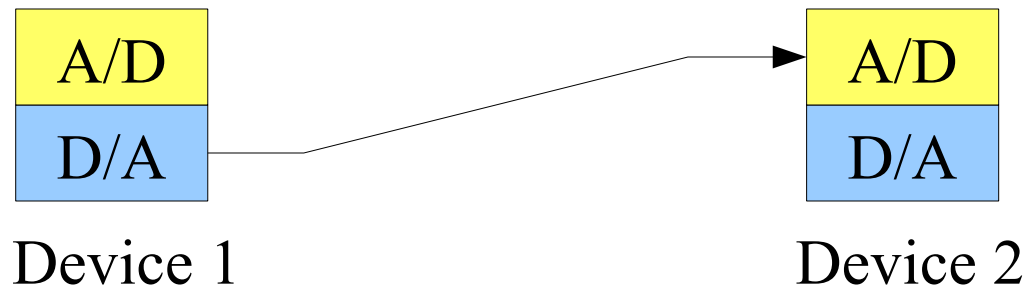


- Allpass chirps can be used for a more robust measure

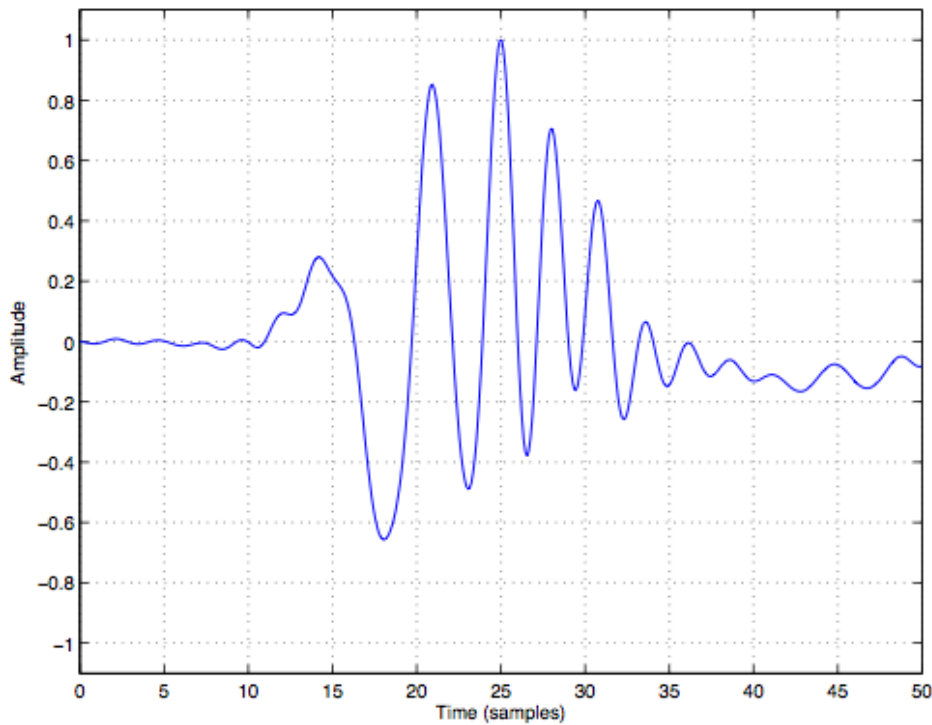


Implicit Estimation

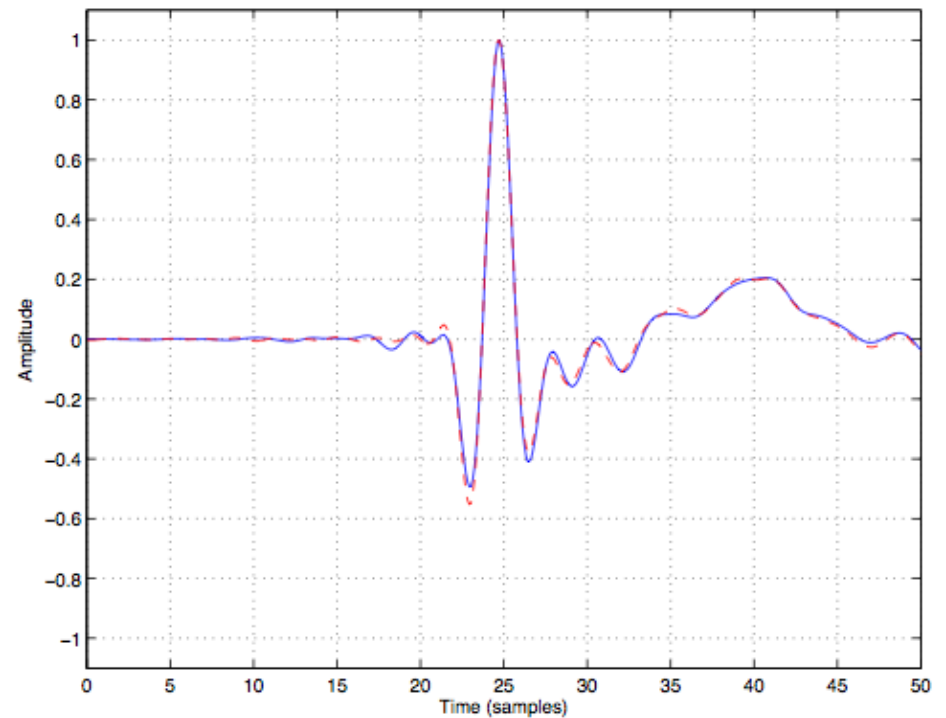
- Record direct loopback of sine sweep or inverse
- Simultaneously estimate drift and resample
- Drift rate *perfectly* estimated, but *unknown*



Compensation Results

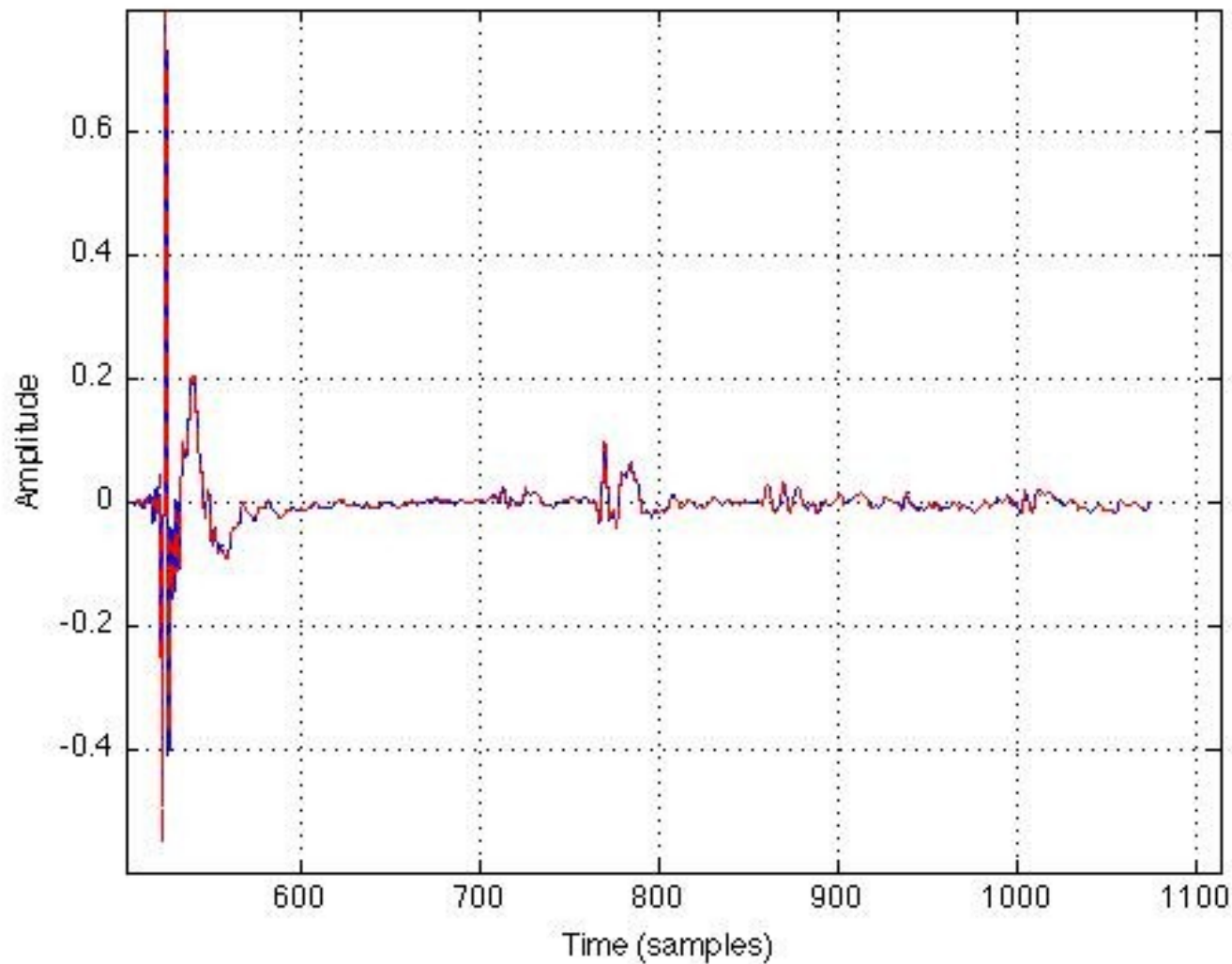


IR direct path with clock drift



IR compensated for drift vs.
reference IR with no drift

Compensation Results



Compensated IR direct path and early reflections

Conclusions

- Clock drift can influence impulse response measurements in various ways
- Unwanted drift imposes allpass filtering on the resulting IR for sinusoidal sweep measurements
- Two methods of compensation are proposed and achieve near perfect compensation
- Post-processing solution for measuring room impulse responses in the presence of clock drift

Acknowledgements & Thank You!

- The Stanford Institute for Creativity and the Arts (SiCa), for the Chavín de Huántar Archaeological Acoustics Project

<https://ccrma.stanford.edu/groups/chavin/>

